



## System Behaviour & Stability

ELEC 3004: **Systems**: Signals & Controls  
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Lecture 8

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### Announcements:



- Assignment 1 is up on Platypus!
  - Due at the end of next week!
  - Next week's tutorial sessions are Q&A meetings

#### Lab 1:

- Extra Session Needed ????
- Space Audit this **Friday Today** & next **Wednesday**
  - Sounds like an ideal time for a **pop-quiz**



Today:

Week	Date	Lecture Title
1	27-Feb	Introduction
	1-Mar	Systems Overview
2	6-Mar	Signals & Signal Models
	8-Mar	System Models
3	13-Mar	Linear Dynamical Systems
	15-Mar	Sampling & Data Acquisition
	20-Mar	Time Domain Analysis of Continuous Time Systems
4	22-Mar	<b>System Behaviour &amp; Stability</b>
5	27-Mar	Signal Representation
	29-Mar	Holiday
6	10-Apr	Frequency Response & Fourier Transform
	12-Apr	Analog Filters
7	17-Apr	IIR Systems
	19-Apr	FIR Systems
8	24-Apr	z-Transform
	26-Apr	Discrete-Time Signals
9	1-May	Discrete-Time Systems
	3-May	Digital Filters
10	8-May	State-Space
	10-May	Controllability & Observability
11	15-May	Introduction to Digital Control
	17-May	Stability of Digital Systems
12	22-May	PID & Computer Control
	24-May	Information Theory & Communications
13	29-May	Applications in Industry
	31-May	Summary and Course Review

### Recall from Last Time:

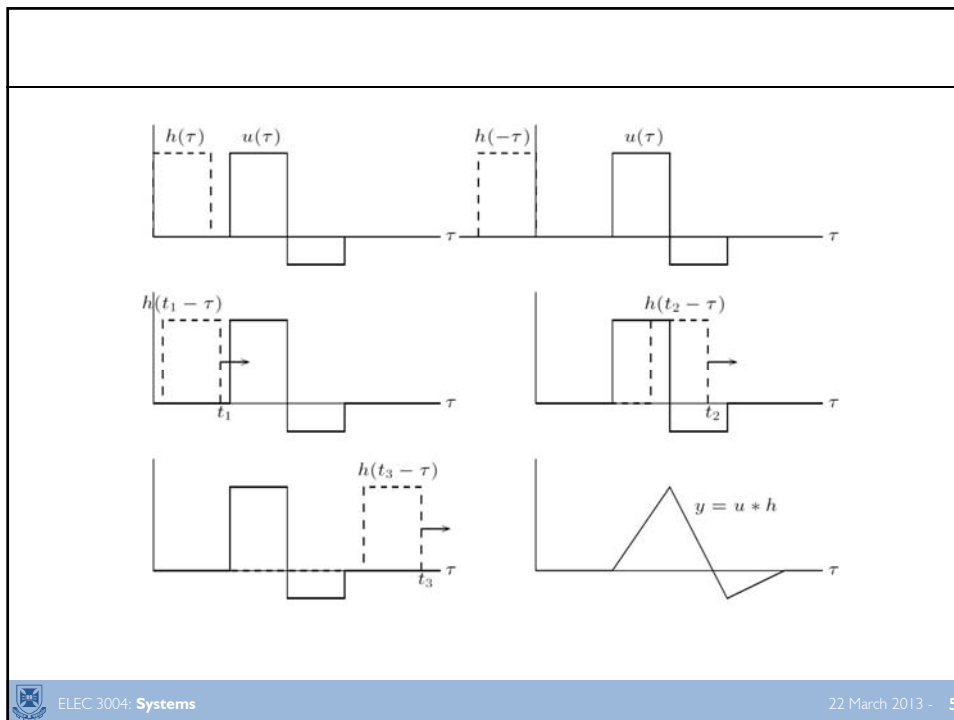
#### Graphical Understanding of Convolution

→ For  $c(\tau) = (f * g)(t) = \int_{-\infty}^{\infty} f(\tau)g(t - \tau) d\tau$  :

1. Keep the function  $f(\tau)$  fixed
2. **Flip** (invert) the function  $g(\tau)$  about the vertical axis ( $\tau=0$ )  
= this is  $g(-\tau)$
3. **Shift** this frame ( $g(-\tau)$ ) along  $\tau$  (horizontal axis) by  $t_0$ .  
= this is  $g(t_0 - \tau)$

→ For  $c(t_0)$ :

4.  $c(t_0)$  = the area under the product of  $f(\tau)$  and  $g(t_0 - \tau)$
5. Repeat this procedure, shifting the frame by different values (positive and negative) to obtain  **$c(t)$  for all values of  $t$** .



## Some Properties of Convolution Systems:

1. convolution systems are **linear**: for all signals  $u_1, u_2$  and all  $\alpha, \beta \in \mathbf{R}$ ,

$$h * (\alpha u_1 + \beta u_2) = \alpha(h * u_1) + \beta(h * u_2)$$

2. convolution systems are **causal**: the output  $y(t)$  at time  $t$  depends only on past inputs  $u(\tau)$ ,  $0 \leq \tau \leq t$

3. convolution systems are **time-invariant**: if we shift the input signal  $u$  over  $T > 0$ , *i.e.*, apply the input

$$\tilde{u}(t) = \begin{cases} 0 & t < T \\ u(t-T) & t \geq 0 \end{cases}$$

to the system, the output is

$$\tilde{y}(t) = \begin{cases} 0 & t < T \\ y(t-T) & t \geq 0 \end{cases}$$

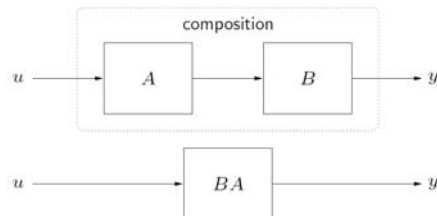
in other words: convolution systems commute with delay



## Some Properties of Convolution Systems:

4. **composition** of convolution systems corresponds to

- multiplication of transfer functions
- convolution of impulse responses



ramifications:

- can manipulate block diagrams with transfer functions as if they were simple gains
- convolution systems commute with each other



## Interpretation of Convolution

$$y(t) = \int_0^t h(\tau)u(t - \tau) d\tau$$

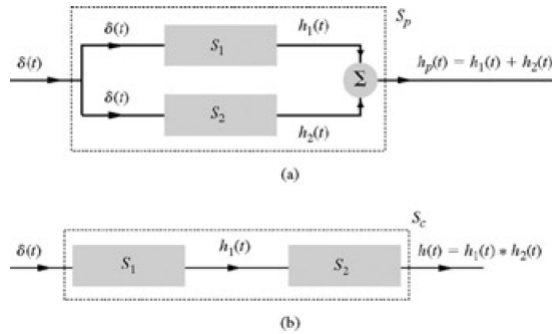
- $y(t)$  is current output;  $u(t - \tau)$  is what the input was  $\tau$  seconds ago
- $h(\tau)$  shows how much current output depends on what input was  $\tau$  seconds ago

for example,

- $h(21)$  big means current output depends quite a bit on what input was, 21sec ago
- if  $h(\tau)$  is small for  $\tau > 3$ , then  $y(t)$  depends mostly on what the input has been over the last 3 seconds
- $h(\tau) \rightarrow 0$  as  $\tau \rightarrow \infty$  means  $y(t)$  depends less and less on remote past input
- LTI + causal = LTI Causal System = "Convolution System"
- Any Convolution System is LTI and causal
- The converse is also true: Any LTI causal system can be represented by a convolution system



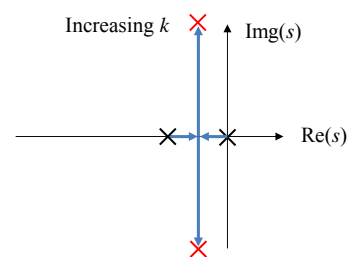
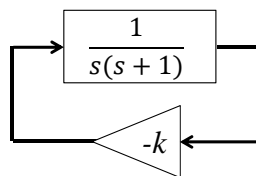
## Interconnected Systems



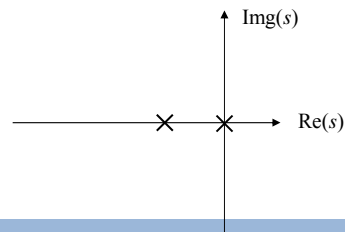
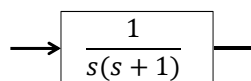
- For more see also Lathi Fig. 2.15

## Recall Root Locus: Pole-Zero Diagram

- Feedback Root Locus:



- Open Loop Root Locus: Pole Zero Diagram



## Stable & Unstable Systems

- Systems can also be classified as **stable** or **unstable** systems
  - Stability can be **internal** or **external**:
    - For external cases: If every bounded input applied at the input terminal results in a bounded output
    - The external stability can be ascertained by measurements at the external terminals (input and output) of the system.
- Stability in the BIBO (bounded-input/bounded-output) sense.



## BIBO (bounded-input/bounded-output)

- For an LTI System:

$$\begin{aligned}y(t) &= h(t) * x(t) \\ &= \int_{-\infty}^{\infty} h(\tau)x(t - \tau) d\tau\end{aligned}$$

Therefore

$$|y(t)| \leq \int_{-\infty}^{\infty} |h(\tau)||x(t - \tau)| d\tau$$

- If  $x(t)$  is bounded:

$$\text{then } |x(t - \tau)| < K_1 < \infty, \text{ and } |y(t)| \leq K_1 \int_{-\infty}^{\infty} |h(\tau)| d\tau$$

- Hence:

$$\int_{-\infty}^{\infty} |h(\tau)| d\tau < \infty$$



# BIBO

