



# Estimation & Course Review

ELEC 3004: **Systems**: Signals & Controls

Dr. Surya Singh  
(with material from Dr. Paul Pounds)

Lecture 25 (The Last!)

[elec3004@itee.uq.edu.au](mailto:elec3004@itee.uq.edu.au)

May 31, 2013

<http://robotics.itee.uq.edu.au/~elec3004/>

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## Today in Linear Systems...

Week	Date	Lecture Title
1	27-Feb	Introduction
	1-Mar	Systems Overview
2	6-Mar	Signals & Signal Models
	8-Mar	System Models
3	13-Mar	Linear Dynamical Systems
	15-Mar	Sampling & Data Acquisition
4	20-Mar	Time Domain Analysis of Continuous Time Systems
	22-Mar	System Behaviour & Stability
5	27-Mar	Signal Representation
	29-Mar	Holiday
6	10-Apr	Frequency Response
	12-Apr	z-Transform
7	17-Apr	Noise & Filtering
	19-Apr	Analog Filters
8	24-Apr	Discrete-Time Signals
	26-Apr	Discrete-Time Systems
9	1-May	Digital Filters & IIR/FIR Systems
	3-May	Fourier Transform & DTFT
10	8-May	Introduction to Digital Control
	10-May	Stability of Digital Systems
11	15-May	PID & Computer Control
	17-May	Applications in Industry
12	22-May	State-Space
	24-May	Controllability & Observability
13	29-May	State-Space: Made Clear
	31-May	Summary and Course Review



</assessable>

## WARNING: NOT ASSESSABLE

- Nothing beyond this point is on the exam.
- Do not pay attention.
- Do not attempt to learn.



## What's Assessable?

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## Announcements:



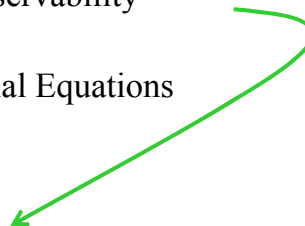
- Practice Final Posted:
  - You're welcome ☺
  - While it's final #3 (real, supp, & prac):
    - It shares **no** questions with either the real or the supp.
  - Exam Review Session: June 7,
    - ➔ It is harder than the real thing!
    - (∴ its primarily to practice the material **not the timing**)
- Problem Set 2 Grading:
  - May 31 (& you get your grades during the semester)
  - June 7 (More time, better feedback??)



## Goals for the Week

Today:

- State-Space: Controllability & Observability
- State-Space: Solution of Differential Equations
- State-Space: Compensator Design



Friday:

- Everything!



# ELEC 3004: A Review

AKA ELEC 3004:  
**What do I need to  
know about \*.\* ???**

## To Review:

### Back to the **Beginning...**Lecture 1 Slide 9

- Systems
- Signal Abstractions
- Signals as Vectors / Systems as Maps
- Linear Systems and Their Properties
- LTI Systems
- Autonomous Linear Dynamical Systems
- Convolution
- FIR & IIR Systems
- Frequency domain
- Fourier Transform (CT)
- Fourier Transform (DT)
- Even and Odd Signals
- Likelihood
- Causality
- Impulse Response
- Root Locus
- Bode Functions
- Left-hand Plane
- Frequency Response
- Discrete Time
- Continuous Time
- Laplace Transformation
- Feedback and Control
- Additional Applications
- Linear Functions
- Linear Algebra Review
- Least Squares
- Least Squares Problems
- Least Squares Applications
- Matrix Decomposition and Linear Algebra
- Regularized Least Squares
- Least-squares
- Least-squares applications
- Orthonormal sets of vectors
- Eigenvectors and diagonalization
- Linear dynamical systems with inputs and outputs
- Symmetric matrices, quadratic forms, matrix norm, and SVD
- Controllability and state transfer
- Observability and state estimation
- And that, of course,  
**Linear Systems are Cool! ☺**



## Review

- What do you think when you see?

$$\ddot{y} + 2\dot{y} + 3y = u$$

- System?
- ODE?
- Linear Algebra?
- Joy?
- Excitement?
- Shock and Awe??

**Linear algebra provides the tools/foundation for working with (linear) differential equations.**



**Linear algebra provides the tools/foundation for working with (linear) differential equations.**

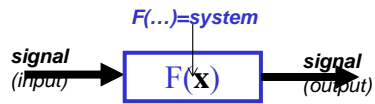
- Signals are vectors. Systems are matrices.

$$\mathbf{y} = \mathbf{F}\mathbf{x} + \mathbf{G}\mathbf{u}$$



# Linear Systems

## Linear Systems in 1-Slide



- Signals Are Vectors
- Systems Are Matrices



Is it Useful?

Yes.

(For example ... Next Year – ELEC/METR 3800)

## It Can Rock Your Boat Gently Down The Stream: IMU Deaduced Reckoning (Navigation)



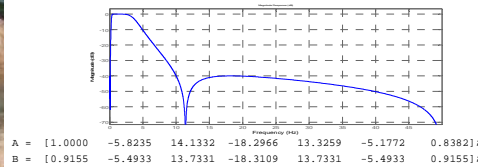
Idea: Integrate your motion (twice for  $\ddot{x} \rightarrow x$  and once for  $\dot{\theta} \rightarrow \theta$ )

Problem:

- (DC) bias in accelerometer  $\rightarrow$  drift

Solution:

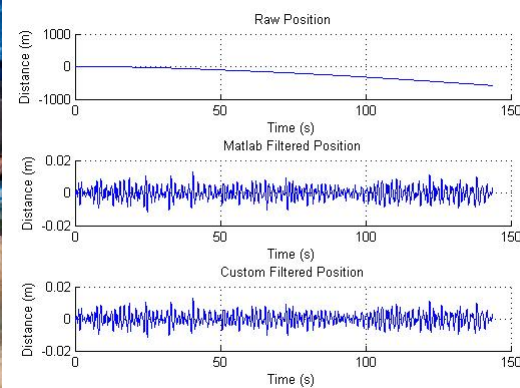
- IIR Bandpass filter (0.1-10 Hz)



## It Can Rock Your Boat Gently Down The Stream: IMU Deaduced Reckoning (Navigation) [2]



Solution:





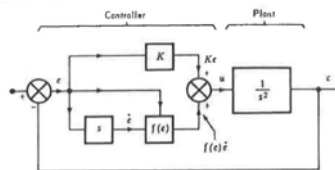
# Extra Material: (For Fun!)

## Poles are Eigenvalues

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### Stability of a 2<sup>nd</sup> order regulator



$$u = Ke + f(e) \dot{e}$$

state equations let  $e = x_1$  and  $\dot{e} = x_2$

$$\dot{x}_1 = x_2$$

$$\dot{x}_2 = -Kx_1 - f(x_1) x_2$$

assume for simplicity that  $K = 1$ .

$$0 = x_2^0$$

$$0 = -x_1^0 - f(x_1^0) x_2^0$$

The Jacobian matrix is

$$A = \begin{bmatrix} 0 & 1 \\ -1 & -f(0) \end{bmatrix}$$

- The linear behavior of the system in the close neighborhood of the origin is described by

$$\dot{x}_1 = x_2$$

$$\dot{x}_2 = -x_1 - f(0) x_2$$

- AND, the characteristic equation is:

$$s[s + f(0)] + 1 = 0$$

with the eigenvalues

$$\lambda_1 = -\frac{1}{2} f(0) + \sqrt{\frac{1}{4} f^2(0) - 1}$$


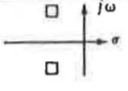

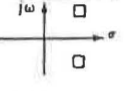

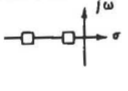

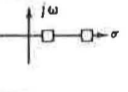

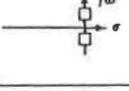

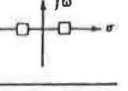
$$\lambda_2 = -\frac{1}{2} f(0) - \sqrt{\frac{1}{4} f^2(0) - 1}$$



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## Various Types of Singularities (2<sup>nd</sup> order systems)

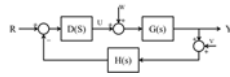
Stable		Unstable	
Trajectory type	Eigenvalues	Trajectory type	Eigenvalues
 Stable focus		 Unstable focus	
 Stable node		 Unstable node	
 Vortex		 Saddle	



## Root Locus Control Design Method (See also Q10. Part F)

## Root Locus Design: “Evan’s Method”

- Imagine “the basic feedback system”:



$$TF(s) = \frac{Y(s)}{R(s)} = \frac{D(s)G(s)}{1 + D(s)G(s)H(s)} = \frac{DG}{1 + DGH}$$

→ Characteristic Equation:  $1 + DGH = 0$

- Put the characteristic equation in Root Locus form:

$$1 + K \cdot L(s) = 0$$

▪ **If** we define  $L(s) = \frac{b(s)}{a(s)}$

▪ **then**  $a(s) + K \cdot b(s) = 0$  and  $L(s) = -\frac{1}{K}$

- Thus a Root Locus is a graph of all possible roots of  $1 + K \cdot L(s) = 0$  and K as the variable parameter

- ∴ This is the solution to the roots of the closed-loop system characteristic equation and thus the closed-loop poles of the system. The root-locus graph may be viewed as a method for interring the **dynamic properties of a system as K changes**.



## The Direct Method of Digital Controls –

**NOT** to be confused with  
**Controller Emulation**  
(e.g., Tustin’s Method)

## Direct Design Method Of Ragazzini (See also: FPW 5.7 pp.216-222)

Start with 3 Discrete Transfer Functions:

- **G(z)**: TF<sup>1</sup> of a **plant + a hold** (e.g., from a ZOH)
- **D(z)**: A **controller** TF to do the job (what we want here)
- **H(z)**: The final desired TF between **R** (reference) and **Y** (output)
- Thus<sup>2</sup>:

$$H(z) = \frac{DG}{1+DG}$$

$$\rightarrow D(z) = \frac{1}{G} \frac{H}{1-H}$$

- This calls for a **D(z)** that will **cancel the plant effects** and that will **add whatever is necessary to give the desired result**. The problem is to discover and implement constraints on **H(z)** so that we do not ask for the impossible.
  - This implies that we **need some constraints** on both **H(z)** and **D(z)**

1: Transfer Function

2: Mental Quiz: What does **1+DG** say about the sign of the feedback (positive or negative)?  
That is, what is the characteristic equation for a system with positive feedback?



## Direct Design Method Of Ragazzini [2]: Design Constraints: I. Causality

- Remember/Recall an Interesting Point:
  - From z-transform theory we know that **if D(z) is causal**, **then** as  $z \rightarrow \infty$  its transfer function is well behaved & **it does not** have a pole at infinity.
- $D(z) = \frac{1}{G} \frac{H}{1-H}$  implies that **if**  $G(z) = 0$  (at  $\infty$ ), **then** **D(z)** would have a pole (at  $\infty$ ) **unless** **H(z)** cancels it.

∴

**H(z)** must have a zero (at  $\infty$ ) of the same order as **G(z)**'s 0s (at  $\infty$ )

→ Which means: **If** there is a lag in the plant (**G(z) starts with  $z^{-l}$** ) **then** causality requires that the delay of **H(z)** is that the closed-loop system must be at least as long a delay of the plant.

(Whoa! It might sound deep, but it's rather intuitive ☺)



## Direct Design Method Of Ragazzini [3]: Design Constraints: II. Stability

- The characteristic equation and the closed loop roots:  

$$1 + D(z)G(z) = 0$$
- Define<sup>3</sup>  $D = \frac{c}{d}$  and  $G = \frac{b}{a} \rightarrow ad + bc = 0$
- Define  $z - \alpha$  as a pole of  $\mathbf{G}(z)$  and a common factor in DG that represents  $\mathbf{D}(z)$  cancelling a pole/zero of  $\mathbf{G}(z)$ .
- **Then** this common factor *remains a factor of the characteristic polynomial*.
- **If** this factor is outside the unit circle, **then** the system is unstable!

$\therefore$

$\mathbf{1-H(z)}$  must contain as zeros  
all the poles of  $\mathbf{G(z)}$  that are outside the unit circle &  
 $\mathbf{H(z)}$  must contain as zeros  
all the zeros of  $\mathbf{G(z)}$  that are outside the unit circle

3: Note the switching of the “alphabetical-ness” of these two fractions



## Direct Design Method Of Ragazzini [4]: Design Constraints: III. Steady State Accuracy

- The error from  $\mathbf{H(z)}$  is given by:  

$$E(z) = R(z)(1 - H(z))$$
- **If** the system is “Type 1” (with a constant velocity/first derivative ( $K_v$ ))
  - **Then**<sup>4</sup>  $E_{ss}^{Step} = 0$  and  $E_{ss}^{Ramp} = 1/K_v$

$\therefore$

$$H(z) = 1$$

&

$$-T_s \left. \frac{dH(z)}{dz} \right|_{z=1} = \frac{1}{K_v} H(z) = 1$$

4:  $E_{ss}$ : steady-state error



## Direct Design Method Of Ragazzini [5]: An Example

- Consider the plant:  $s^2 + s + 1 = 0$   
With  $T_s=1 \rightarrow z$ -Transform:  $z^2 + 0.786z + 0.368=0$
- Let's design this **system** such that
  - $K_v = 1$
  - Poles at the roots of the plant equation & additional poles as needed

$$\rightarrow H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2} + b_3 z^{-3} + \dots}{1 - 0.786z^{-1} + 0.368 z^{-2}}$$

I. Causality:  $H(z)|_{z=\infty} = 0 \rightarrow b_0 = 0$

II. Stability: All poles/zeros of  $G(z)$  are in the unit circle  
 – except for  $b_0$ , which is taken care of by  $b_0 = [Const] = 0$

III. Tracking:

- $H(1) = b_1 + b_2 + b_3 + \dots = 1 \cdot (1 - 0.786 + 0.368)$  &
- $-\{1\} \frac{dH(z)}{d(z^{-1})} \Big|_{z=1} = \frac{1}{\{1\}} \rightarrow \frac{b_1 + 2b_2 + 3b_3 + \dots - [-0.05014]}{(1 - 0.786 + 0.368)}$  (note the  $z^{-1}$ )
- Truncate the number of unknowns to 2 “zeros” ... thus solve for  $b_1$  and  $b_2$  (& set  $b_3, b_4, \dots = 0$ )

$$\therefore H(z) = \frac{b_1 z + b_2}{z^2 - 0.786z + 0.368} \rightarrow D(z) = \frac{(z-1)(z-0.9048)(0.5321)}{(0.04837)(z+0.9672)} \frac{(z-0.07932)}{(z-1)(z-0.4180)}$$

$$= 13.07 \frac{(z-0.9048)(z-0.07932)}{(z+0.9672)(z-0.4180)}$$



# Digital Controls (Magic PID Made Easy Equations)

## Implementation of Digital PID Controllers

We will consider the PID controller with an  $s$ -domain transfer function

$$\frac{U(s)}{X(s)} = G_c(s) = K_P + \frac{K_I}{s} + K_D s. \quad (13.54)$$

We can determine a digital implementation of this controller by using a discrete approximation for the derivative and integration. For the time derivative, we use the **backward difference rule**

$$u(kT) = \left. \frac{dx}{dt} \right|_{t=kT} = \frac{1}{T} (x(kT) - x[(k-1)T]). \quad (13.55)$$

The  $z$ -transform of Equation (13.55) is then

$$U(z) = \frac{1 - z^{-1}}{T} X(z) = \frac{z - 1}{Tz} X(z).$$

The integration of  $x(t)$  can be represented by the **forward-rectangular integration** at  $t = kT$  as

$$u(kT) = u[(k-1)T] + Tx(kT), \quad (13.56)$$

Source: Dorf & Bishop, Modern Control Systems, §13.9, pp. 1030-1



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## Implementation of Digital PID Controllers (2)

where  $u(kT)$  is the output of the integrator at  $t = kT$ . The  $z$ -transform of Equation (13.56) is

$$U(z) = z^{-1}U(z) + TX(z),$$

and the transfer function is then

$$\frac{U(z)}{X(z)} = \frac{Tz}{z - 1}.$$

Hence, the  $z$ -domain transfer function of the **PID controller** is

$$G_c(z) = K_P + \frac{K_I T z}{z - 1} + K_D \frac{z - 1}{Tz}. \quad (13.57)$$

The complete difference equation algorithm that provides the PID controller is obtained by adding the three terms to obtain [we use  $x(kT) = x(k)$ ]

$$\begin{aligned} u(k) &= K_P x(k) + K_I [u(k-1) + Tx(k)] + (K_D/T)[x(k) - x(k-1)] \\ &= [K_P + K_I T + (K_D/T)]x(k) - K_D T x(k-1) + K_I u(k-1). \end{aligned} \quad (13.58)$$

Equation (13.58) can be implemented using a digital computer or microprocessor. Of course, we can obtain a PI or PD controller by setting an appropriate gain equal to zero.

Source: Dorf & Bishop, Modern Control Systems, §13.9, pp. 1030-1



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## Implementation of Digital PID Controllers (2)

- In FPW Terminology (FPW 5.8.4 p. 224)

### 5.8.4 PID Control

Combining all the above yields the PID controller

$$D(z) = K_p \left( 1 + \frac{Tz}{T_I(z-1)} + \frac{T_D(z-1)}{Tz} \right). \quad (5.61)$$

This form of control law is able satisfactorily to meet the specifications for a large portion of control problems and is therefore packaged commercially and sold for general use. The user simply has to determine the best values of  $K_p$ ,  $T_D$ , and  $T_I$ .



# Now, What's Next?



## What's Next?

### Research:

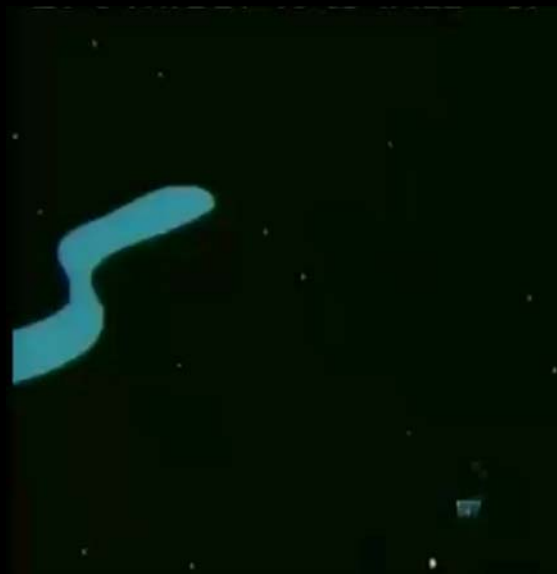
- Thesis Projects:
  - Signal Processing (Eulerian Video Magnification)
  - Digital Control (OpenROVs)
  - Robotics
  - More!

### Courses

- ELEC4620:  
Digital Signal Processing
- ELEC4630:  
Image Processing &  
Computer Vision
- **METR 4202:**  
Advanced Control & Robotics
- CSSE4010:  
Digital System Design



Today's Lecture is Brought To You By the Number 5



## SECATs: One more Systems Example

- Is ELEC 3004 Linear?
- ELEC Controllability?
- Is it / the instructor Stable?




## SECATs: Let's look back at the topic list from Lecture 1

The course is has a huge mandate:

- It is really  $3 \cdot \frac{1}{2}$  courses in one !
  - Linear Systems
  - Signal Processing
  - Controls & Digital Controls
- $\therefore$  It is **b r o a d !!**
- There is a logic to it
  - They share the same mathematical nature (poles & zeros)
  - The math is common to more than just circuits!



## Lots of Stuff To Cover...

✓ Systems	✓ Frequency Response	✓ Symmetric matrices, quadratic forms, matrix norm, and SVD
✓ Signal Abstractions	✓ Discrete Time	✓ Controllability and state transfer
✓ Signals as Vectors / Systems as Maps	✓ Continuous Time	✓ Observability and state estimation
✓ Linear Systems and Their Properties	✓ Laplace Transformation	• And that, of course,
✓ LTI Systems	✓ Feedback and Control	<b>Linear Systems are Cool! ☺</b> 
✓ Autonomous Linear Dynamical Systems	✓ Additional Applications	
✓ Convolution	✓ Linear Functions	
✓ FIR & IIR Systems	✓ Linear Algebra Review	
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✓ Root Locus	✓ Least-squares applications	
✓ Bode Functions	✗ Orthonormal sets of vectors	
	✓ Eigenvectors and diagonalization	
✓ Left-hand Plane	✓ Linear dynamical systems with inputs and outputs	



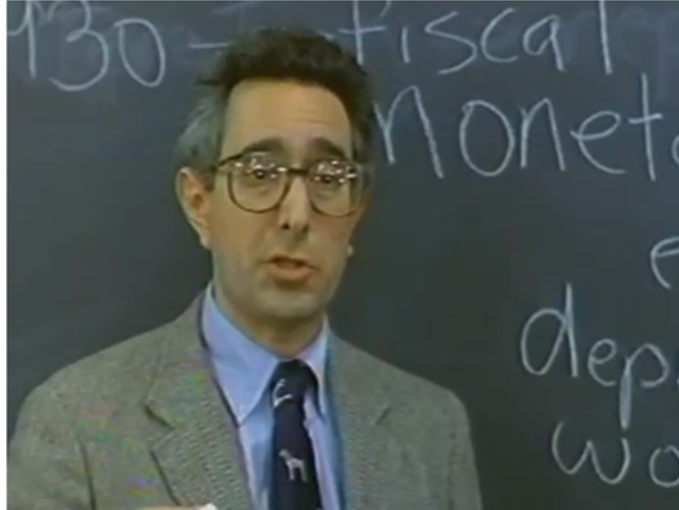
## Yes, this is Hard! Why?

- Breath
  - Books, books, everywhere, yet we're all on Wikipedia!!
  - Authors tend to be “too generalizable”
- Assumptions:
  - Numerous conditions that need to be remembered
- Tacit Details:
  - The need for examples (but these are few and always seem the same)
- Time consuming



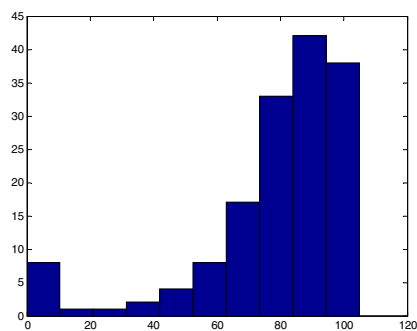
## “4” Is Average

- What is a 3?



## Grades: Better than you think!

- Problem Set 1
- Extra Credit



	Lab 1	Lab 2	Lab 3	Lab 4
Mean	3.683486239	3.305084746	3.65	3.229167
Median	4	3	4	3
St. Dev	0.853112355	0.764780622	0.47697	0.62047

- Quiz 1, 2, & 3:

	Quiz 1	Quiz 2	Quiz 3
Mean	7.637868	7.535433	4.706522
Median	7.5	8	5
St. Dev	1.651598	2.056468	1.115971



## SECaTs: Some Lessons in the Works for Next Year

- I shall only use my own slides
  - Less is more!
    - Smaller assignments
    - More time for Examples
  - Revised Platypus (using TinyMCE)
  - Better organization
    - Better (more structured) tutorials
    - More examples!!
    - I get that. But, we've come a long way
- To make this happen I need your support!



## Next Time in Linear Systems ....

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- We're at the End. It's (the) final!
- Thank you folks!



# Now Finally Some Philosophy (I am a Dr of it!!!) Systems: Signals, Controls...A Fundamental Yearn!

*If you want to build a ship, don't drum up the men to gather wood, divide the work and give orders. **Instead, teach them to yearn for the vast and endless sea.***

Antoine de Saint-Exupery, "The Wisdom of the Sands"

© National Geographic. Mount Everest at night  
(the lights along the apex are the headlamps of other mountaineers)

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