



Frequency Response

ELEC 3004: Systems: Signals & Controls
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Lecture 10

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Today

Week	Date	Lecture Title
	27-Feb	Introduction
1	1-Mar	Systems Overview
2	6-Mar	Signals & Signal Models
	8-Mar	System Models
3	13-Mar	Linear Dynamical Systems
	15-Mar	Sampling & Data Acquisition
4	20-Mar	Time Domain Analysis of Continuous Time Systems
	22-Mar	System Behaviour & Stability
5	27-Mar	Signal Representation
	29-Mar	Holiday
6	10-Apr	Frequency Response & Fourier Transform
	12-Apr	Analog Filters
7	17-Apr	IIR Systems
	19-Apr	FIR Systems
8	24-Apr	z-Transform
	26-Apr	Discrete-Time Signals
9	1-May	Discrete-Time Systems
	3-May	Digital Filters
10	8-May	State-Space
	10-May	Controllability & Observability
11	15-May	Introduction to Digital Control
	17-May	Stability of Digital Systems
12	22-May	PID & Computer Control
	24-May	Information Theory & Communications
13	29-May	Applications in Industry
	31-May	Summary and Course Review



Cool Signal Share: Eulerian Video Magnification for Revealing Subtle Changes in the World



Eulerian Video Magnification for Revealing Subtle Changes in the World

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John Guttag¹ Frédo Durand¹ William T. Freeman¹

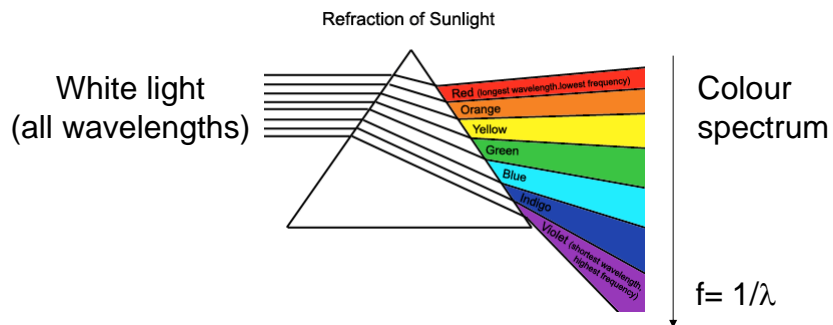
¹MIT CSAIL ²Quanta Research Cambridge, Inc.

Announcements:

- Assignment 1 Solutions:
 - Posted
 - Please try to get the peer review marks in by **April 16** at 11:59pm



Signals Processing: Seeing The Light



Think of a Signal Processing like a prism:
“Destructs a source signal into its constituent frequencies”



Fourier Series

- Any finite power, periodic, signal $x(t)$
 - period T
- can be represented as (∞) summation of
 - sine and cosine waves
- Called: Trigonometrical Fourier Series

$$x(t) = \frac{A_0}{2} + \sum_{n=1}^{\infty} A_n \cos(n\omega_0 t) + B_n \sin(n\omega_0 t)$$

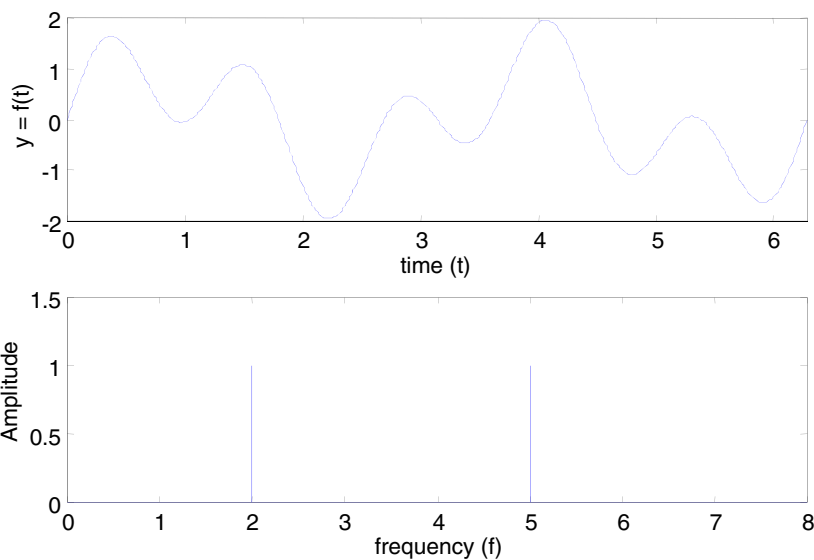
- Fundamental frequency $\omega_0 = 2\pi/T$ rad/s or $1/T$ Hz
- DC (average) value $A_0/2$



Transform Analysis

- Signal measured (or known) as a function of an independent variable
 - e.g., time: $y = f(t)$
- However, this independent variable may not be the most appropriate/informative
 - e.g., frequency: $Y = f(w)$
- Therefore, need to transform from one domain to the other
 - e.g., time \leftrightarrow frequency
 - As used by the human ear (and eye)

Signal processing uses Fourier, Laplace, & z transforms etc



Frequency representation (spectrum) shows signal contains:

- 2Hz and 5Hz components (sinewaves) of equal amplitude



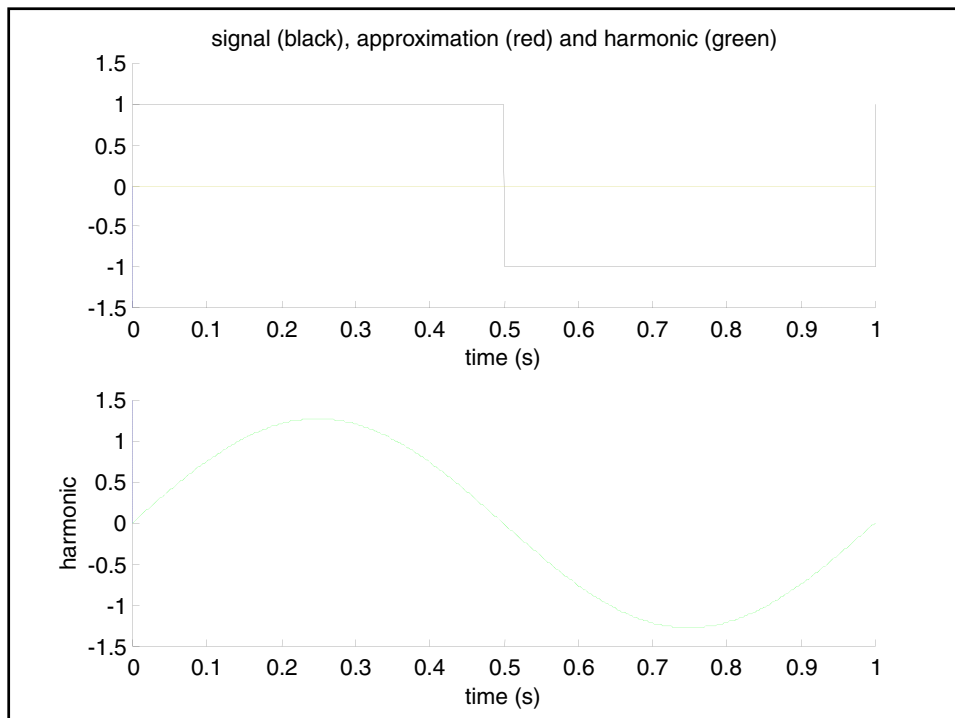
Fourier Series Coefficients

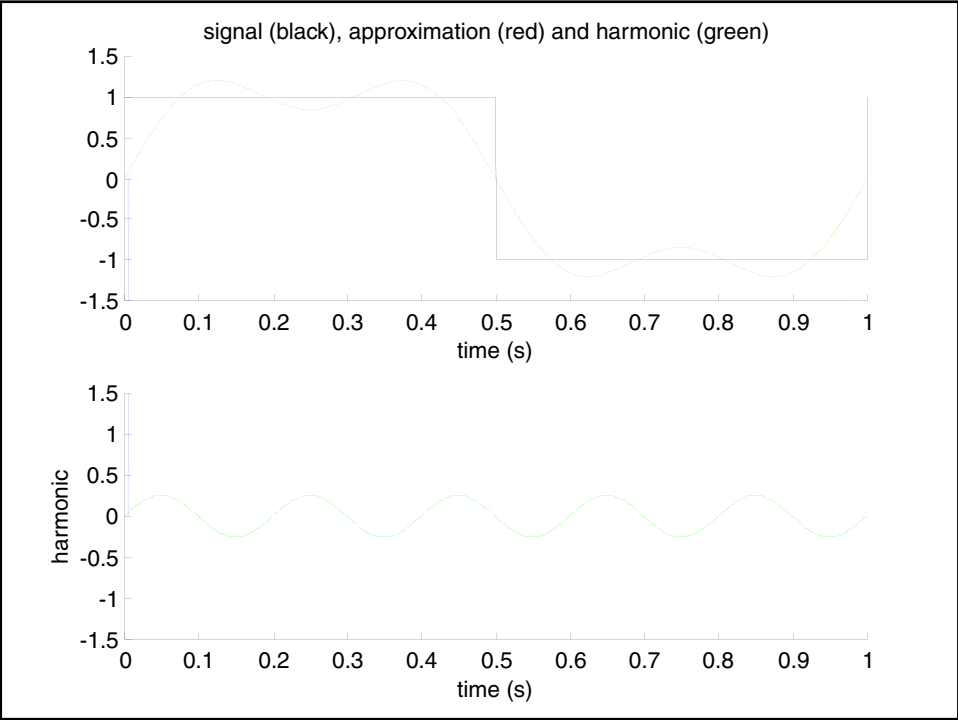
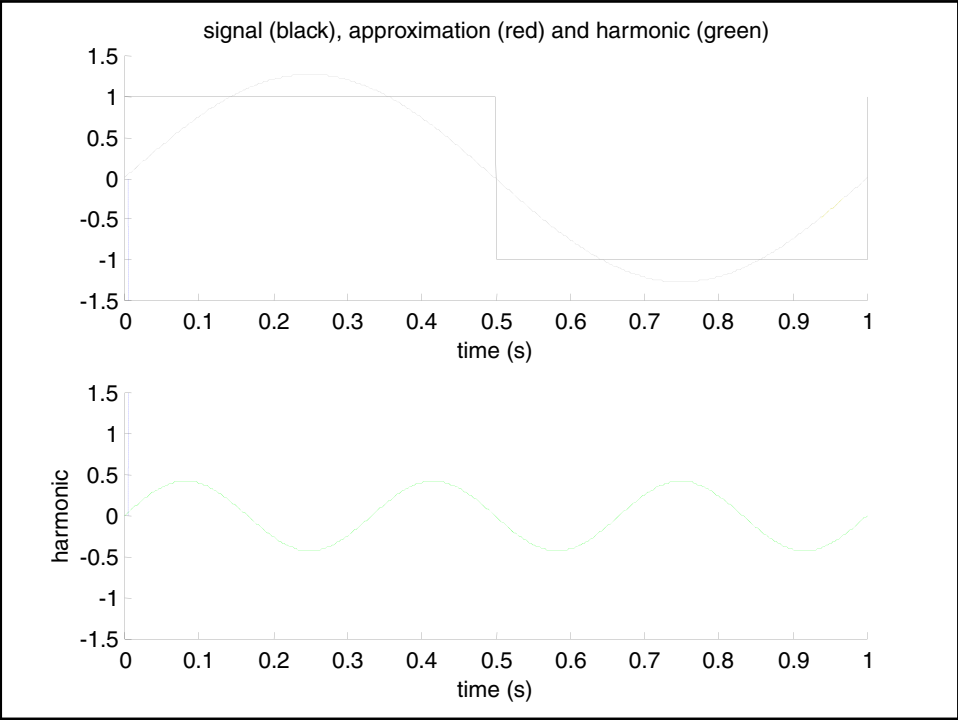
- A_n & B_n calculated from the signal, $x(t)$
 - called: Fourier coefficients

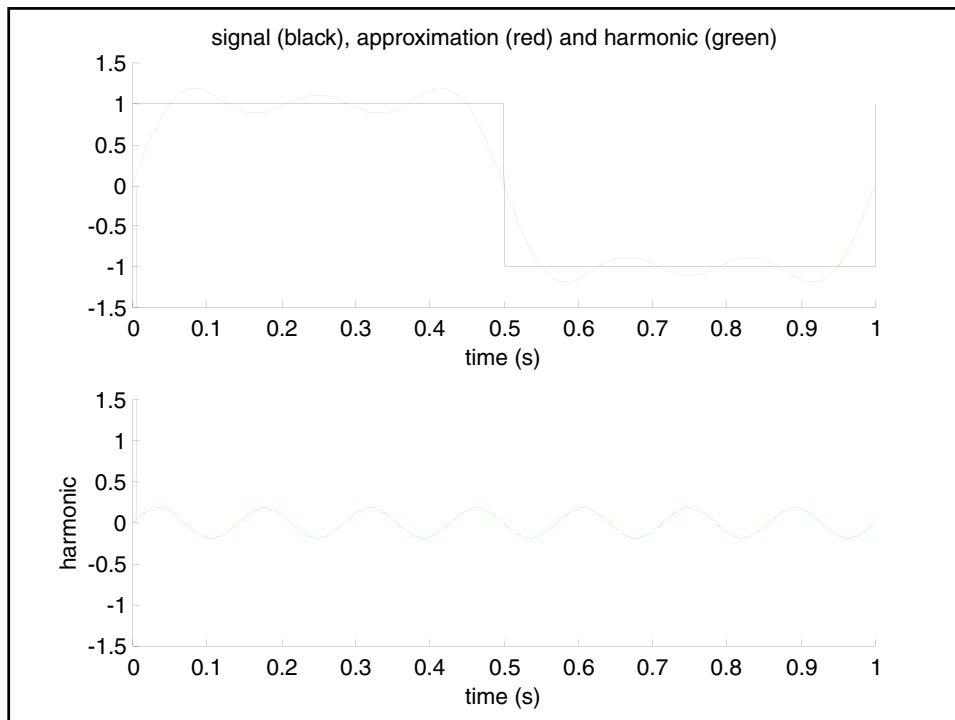
$$A_n = \frac{2}{T} \int_{-T/2}^{T/2} x(t) \cos(n\omega_0 t) dt \quad n = 0, 1, 2, \dots$$

$$B_n = \frac{2}{T} \int_{-T/2}^{T/2} x(t) \sin(n\omega_0 t) dt \quad n = 1, 2, 3, \dots$$

Note: Limits of integration can vary, provided they cover one period







Fourier Series Coefficients

- Approximation with 1st, 3rd, 5th, & 7th Harmonics added, note:
 - ‘Ringing’ on edges due to series truncation
 - Often referred to as Gibb’s phenomenon
- Fourier series converges to original signal if
 - Dirichlet conditions satisfied
 - Closer approximation with more harmonics

Dirichlet Conditions

- For Fourier series to converge, $f(t)$ must be
- defined & single valued
- continuous and have a finite number of finite discontinuities within a periodic interval, and
- piecewise continuous in periodic interval, as must $f'(t)$



Example: Square wave

$$x(t) = \begin{cases} 1, & 0 < t < 1; \\ -1, & 1 < t < 2; \\ x(t+2), & \leftarrow \text{periodic! i.e., } x(t+2) = x(t) \end{cases}$$

$$A_n = \int_0^2 x(t) \cos(n\pi t) dt = \int_0^1 \cos(n\pi t) dt - \int_1^2 \cos(n\pi t) dt$$

$$A_n = \left[\frac{-\sin(n\pi t)}{n\pi} \right]_0^1 - \left[\frac{-\sin(n\pi t)}{n\pi} \right]_1^2 = 0 \quad \begin{array}{l} \text{No cos terms as } \sin(n\pi) = 0 \forall n \\ x(t) \text{ has odd symmetry} \end{array}$$

$$B_n = \int_0^2 x(t) \sin(n\pi t) dt = \int_0^1 \sin(n\pi t) dt - \int_1^2 \sin(n\pi t) dt$$

$$B_n = \left[\frac{-\cos(n\pi t)}{n\pi} \right]_0^1 - \left[\frac{-\cos(n\pi t)}{n\pi} \right]_1^2 = -\frac{\cos(n\pi)}{n\pi} + \frac{1}{n\pi} + \frac{1}{n\pi} - \frac{\cos(n\pi)}{n\pi}$$

$\swarrow \cos(2n\pi) = 1 \forall n$

$$B_n = \frac{2}{n\pi} (1 - \cos(n\pi)) \quad \text{Sin terms only}$$



Example: Square wave

Therefore, Trigonometric Fourier series is,

$$x(t) = \sum_{n=1}^{\infty} \frac{2}{n\pi} (1 - \cos(n\pi)) \sin(n\pi t)$$

Expanding the terms gives,

$$\begin{aligned} x(t) = & \frac{4}{\pi} \sin(\pi t) && \text{(fundamental)} \\ & + 0 && \text{(second harmonic)} \\ & + \frac{4}{3\pi} \sin(3\pi t) && \text{(third harmonic)} \\ & + 0 && \text{(fourth harmonic)} \\ & + \frac{4}{5\pi} \sin(5\pi t) && \text{(fifth harmonic)} \\ & + \text{etc} \end{aligned}$$

- Only odd harmonics;
- In proportion
1, 1/3, 1/5, 1/7, ...
- Higher harmonics contribute less;
- Therefore, converges



Complex Fourier Series (CFS)

- Also called Exponential Fourier series
- FS as a Complex phasor summation
- As it uses Euler's relation

$$A \exp(j\omega_0 t) = A \cos(\omega_0 t) + jA \sin(\omega_0 t)$$

which implies,

$$\cos(n\omega_0 t) = \frac{\exp(jn\omega_0 t) + \exp(-jn\omega_0 t)}{2}$$

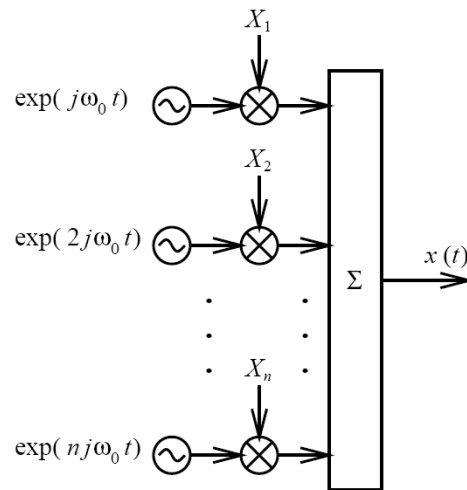
$$\sin(n\omega_0 t) = \frac{\exp(jn\omega_0 t) - \exp(-jn\omega_0 t)}{2j}$$

$$x(t) = \sum_{n=-\infty}^{+\infty} X_n \exp(jn\omega_0 t)$$

Where X_n are the CFS coefficients



Complex Phasor Summation



Complex Fourier Coefficients

- Again, X_n calculated from $x(t)$
- Only one set of coefficients, X_n
 - but, generally they are complex

$$X_n = \frac{1}{T} \int_{-T/2}^{+T/2} x(t) \exp(-jn\omega_0 t) dt$$

Remember: fundamental $\omega_0 = 2\pi/T$!



Relationships

- There is a simple relationship between
 - trigonometrical and
 - complex Fourier coefficients,

$$X_0 = \frac{A_0}{2}$$

$$X_n = \begin{cases} \frac{A_n - jB_n}{2}, & n > 0; \\ \frac{A_n + jB_n}{2}, & n < 0. \end{cases}$$

Constrained to be symmetrical, i.e., complex conjugate

$$X_{-n} = X_n^*$$

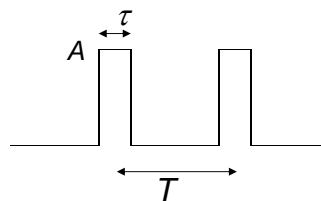
Therefore, can calculate simplest form and convert



Example: Complex FS

- Consider the pulse train signal
- Has complex Fourier series:

$$x(t) = \begin{cases} A, & 0 \leq |t| \leq \frac{\tau}{2}; \\ 0, & \frac{\tau}{2} < |t| \leq T; \\ x(t+T). \end{cases}$$



$$X_n = \frac{1}{T} \int_{-T/2}^{T/2} x(t) \exp(-jn\omega_0 t) dt = \frac{1}{T} \int_{-\tau/2}^{\tau/2} A \exp(-jn\omega_0 t) dt$$

$$= \frac{-A\tau}{jn\omega_0 T \tau} \left[\exp\left(\frac{-jn\omega_0 \tau}{2}\right) - \exp\left(\frac{jn\omega_0 \tau}{2}\right) \right]$$

Note: x by τ ...

Note: n is the ind. variable



Example: Complex FS

- Which using Euler's identity reduces to:

$$X_n = \frac{A\tau}{T} \frac{\sin(n\omega_0\tau/2)}{n\omega_0\tau/2} = \frac{A\tau}{T} \text{sa}(n\omega_0\tau/2)$$

$$\omega_0 = \frac{2\pi}{T}$$

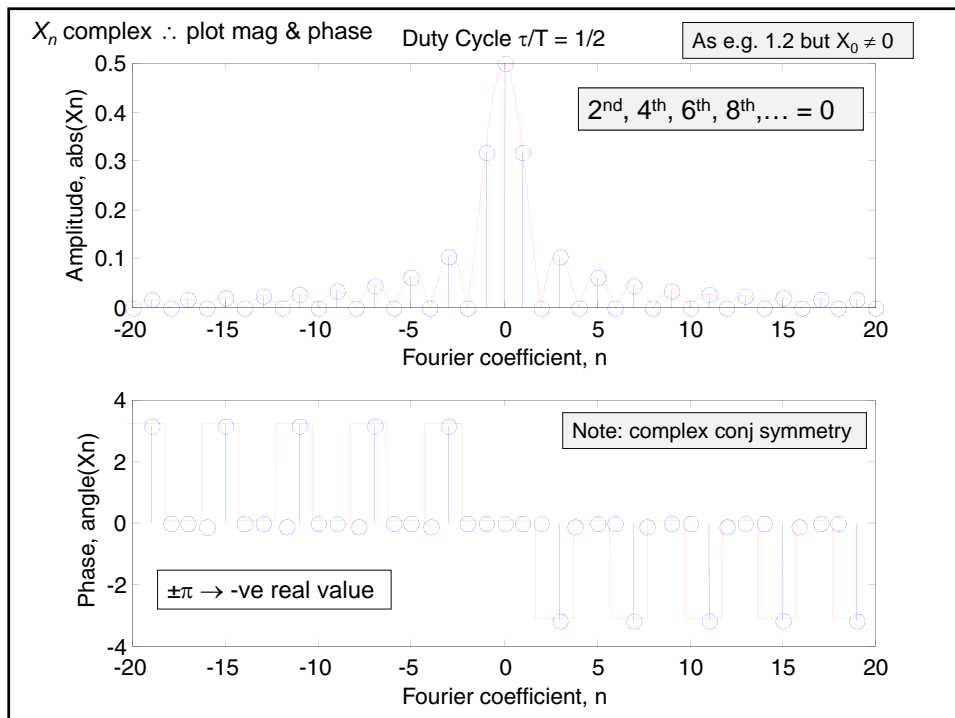
Note: letting $\theta = \frac{n\omega_0\tau}{2}$

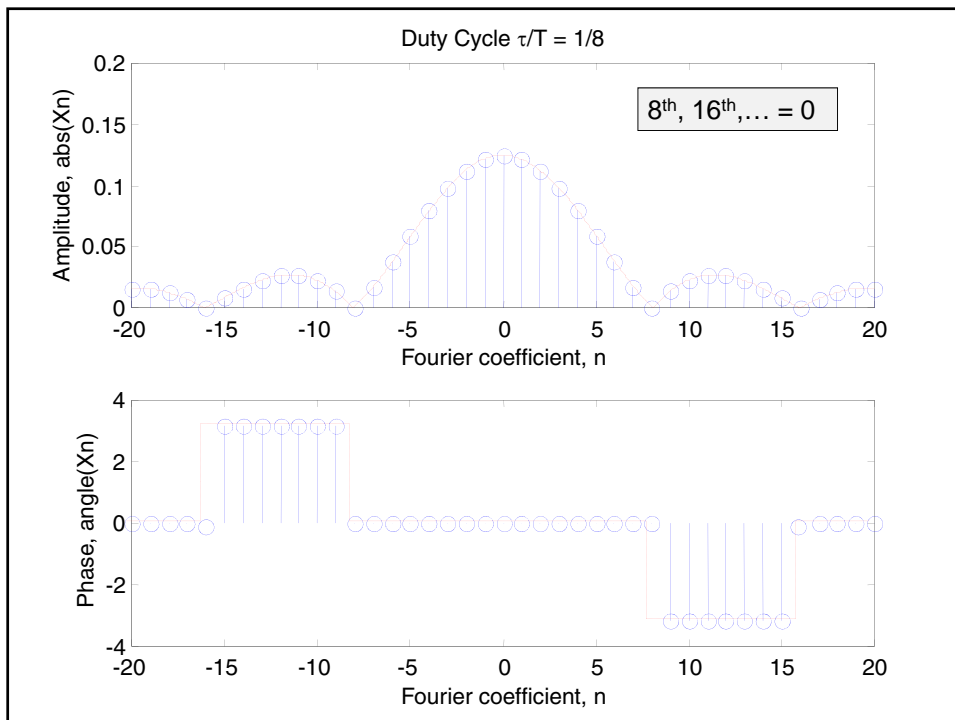
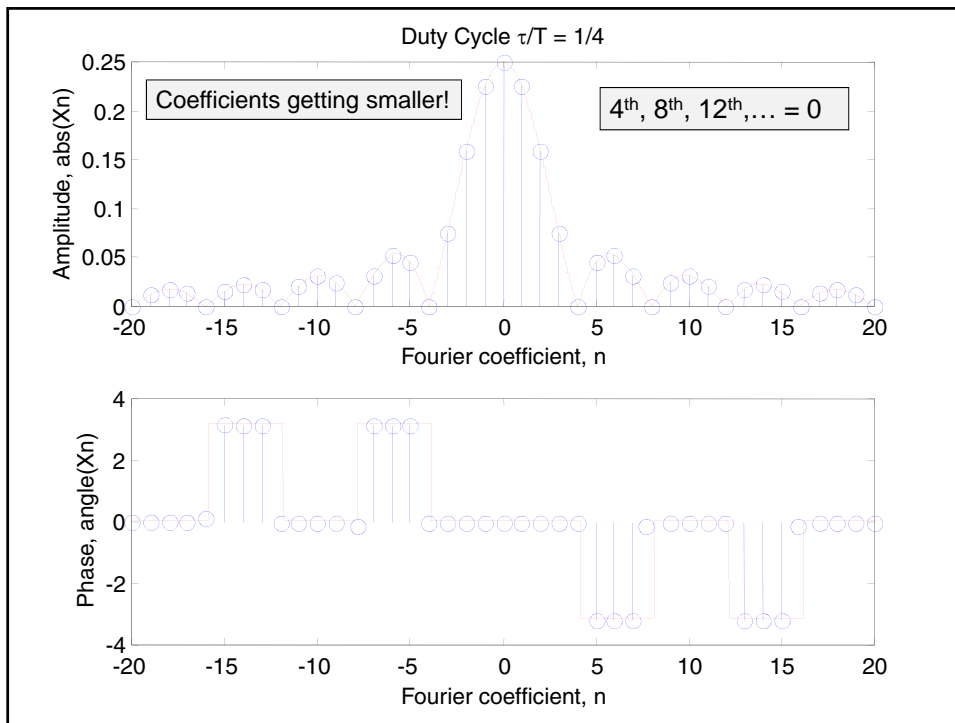
$$\exp(-j\theta) - \exp(j\theta)$$

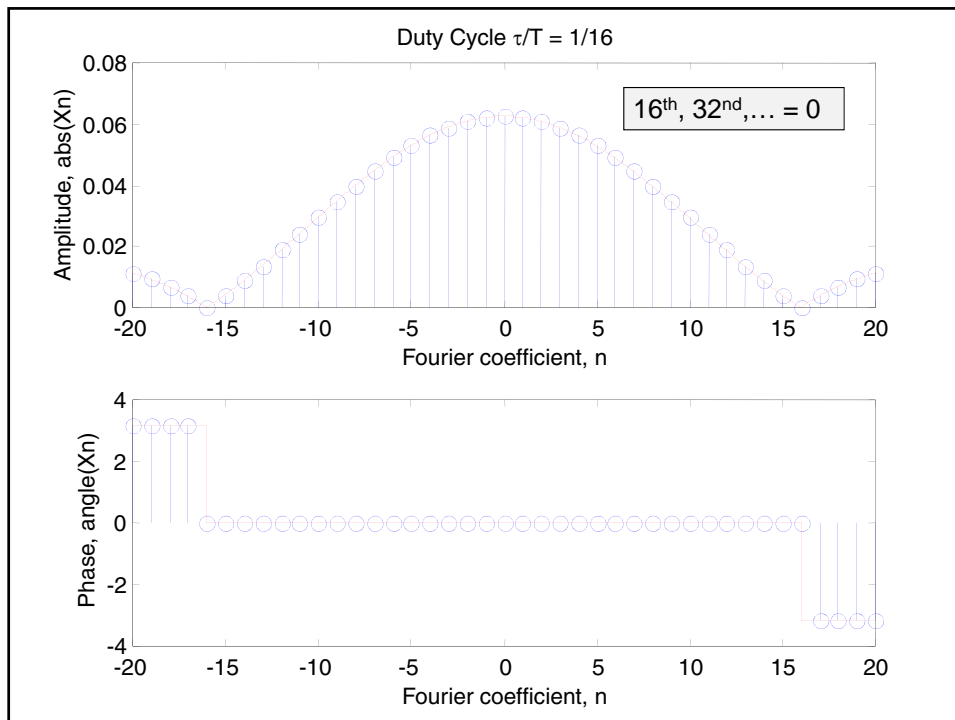
$$= \cos(-\theta) + j\sin(-\theta) - (\cos(\theta) + j\sin(\theta))$$

$$= \cos(\theta) - j\sin(\theta) - \cos(\theta) - j\sin(\theta) = -2j\sin(\theta)$$

Note:
 $\cos(-\theta) = \cos(\theta)$: even
 $\sin(-\theta) = -\sin(\theta)$: odd







Complex FS of Pulse Train

- Amplitude spectrum has ‘sinc’ (‘sa’) envelope
 - Amplitude reduces as duty cycle decreases
 - DC coefficient X_0 ($n=0$) present
 - compare to first example
 - Duty cycle τ/T : $X_M \tau/T = 0$ ($M = 1, 2, 3, \dots$)
 - Even symmetry: $|X_{-n}| = |X_n|$
 - For real (not complex) signals (all we shall consider)
 - Often only plot positive frequency, e.g., Matlab
- Phase spectrum
 - Odd symmetry: $\angle X_{-n} = -\angle X_n$

Complex conjugate (Hermitian) symmetry is general property for real $x(t)$

Interpretation of Fourier Series

- Represents periodic signals ($T = 2\pi/\omega_0$)
 - as sum of cosine waves: “cosine series”
 - at harmonic frequencies $0, \omega_0, 2\omega_0, 3\omega_0, \dots$
 - $|X_n|$ is half amplitude of n th harmonic
 - $\angle X_n$ is phase shift of n th harmonic
- Distribution with harmonic number of
 - both amplitude & phase
 - called a frequency spectrum (discrete)

$$x(t) = X_0 + \sum_{n=1}^{\infty} 2 |X_n| \cos(n\omega_0 t + \angle X_n)$$

i.e., Harmonics only



Orthogonal Expansions

- Trigonometrical and Complex FS both
 - Orthogonal expansions
- Because harmonically related
 - sines, cosines, and complex phasors are all orthogonal
 - Product of $f_1(t)$ & $f_2(t)$ integrates to zero over 1 period

$$\frac{1}{T} \int_{-T/2}^{T/2} \exp(jn\omega_0 t) \exp^*(jm\omega_0 t) dt = \begin{cases} 1, & \text{if } m = n; \\ 0, & \text{if } m \neq n. \end{cases}$$

This means we can calculate each X_n independently
(instead of solving n simultaneous equations)



Example: Calculate Power in x(t)

A simple periodic signal $x(t)$ where:

$$x(t) = a_1 \sin(\omega_0 t) + a_2 \sin(2\omega_0 t)$$

The power:

$$\begin{aligned} P &= \frac{1}{T} \int_0^T x^2(t) dt \\ &= \frac{1}{T} \int_0^T a_1^2 \sin^2(\omega_0 t) dt && \text{Orthogonal} \rightarrow 0 \\ &\quad + \frac{2}{T} \int_0^T a_1 a_2 \sin(\omega_0 t) \sin(2\omega_0 t) dt \\ &\quad + \frac{1}{T} \int_0^T a_2^2 \sin^2(2\omega_0 t) dt \\ &= \frac{1}{T} \int_0^T a_1^2 \sin^2(\omega_0 t) dt + \frac{1}{T} \int_0^T a_2^2 \sin^2(2\omega_0 t) dt \\ &= \frac{a_1^2}{2} + \frac{a_2^2}{2} && \text{Total = sum of power in 2 sine waves} \end{aligned}$$



Parseval's Theorem

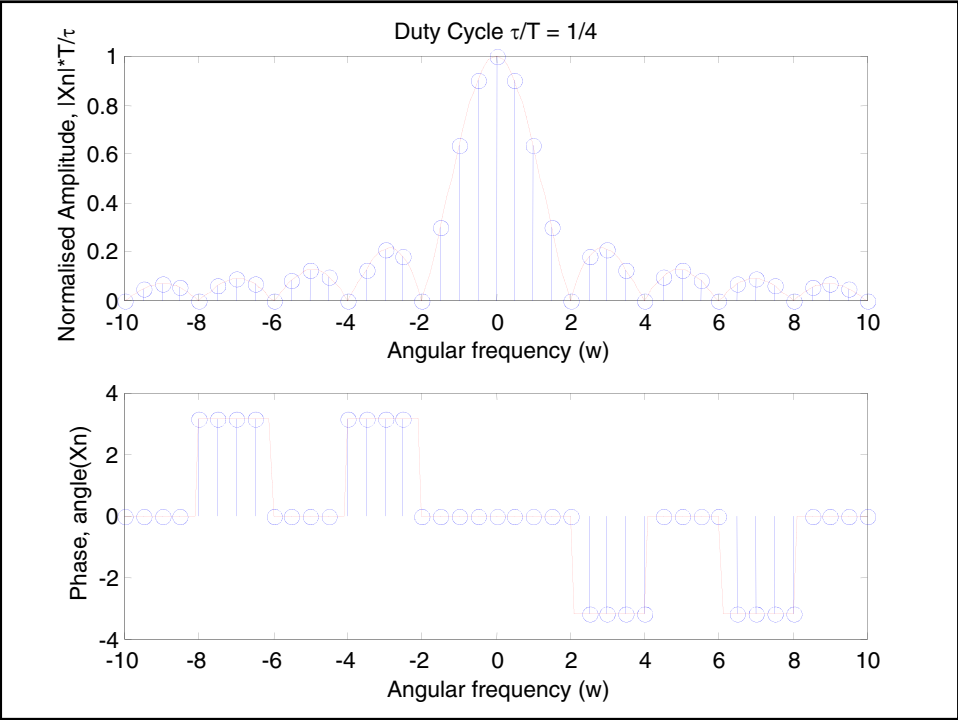
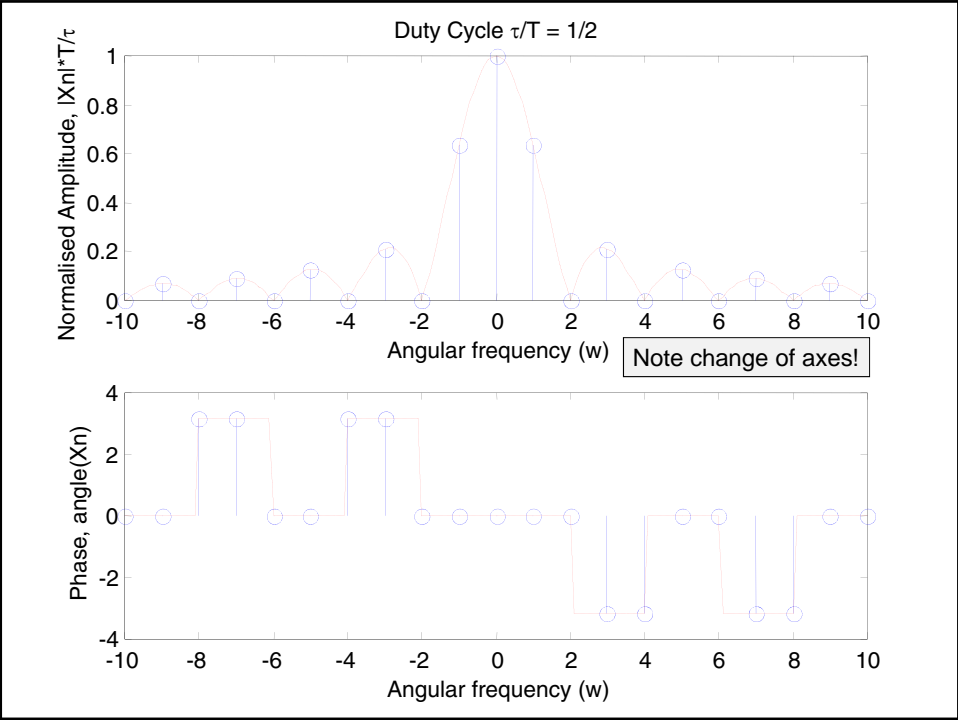
- Direct consequence of orthogonality
- Calculate power of a signal either
 - in time domain, i.e., from $x(t)$, or
 - in frequency domain, i.e., from X_n

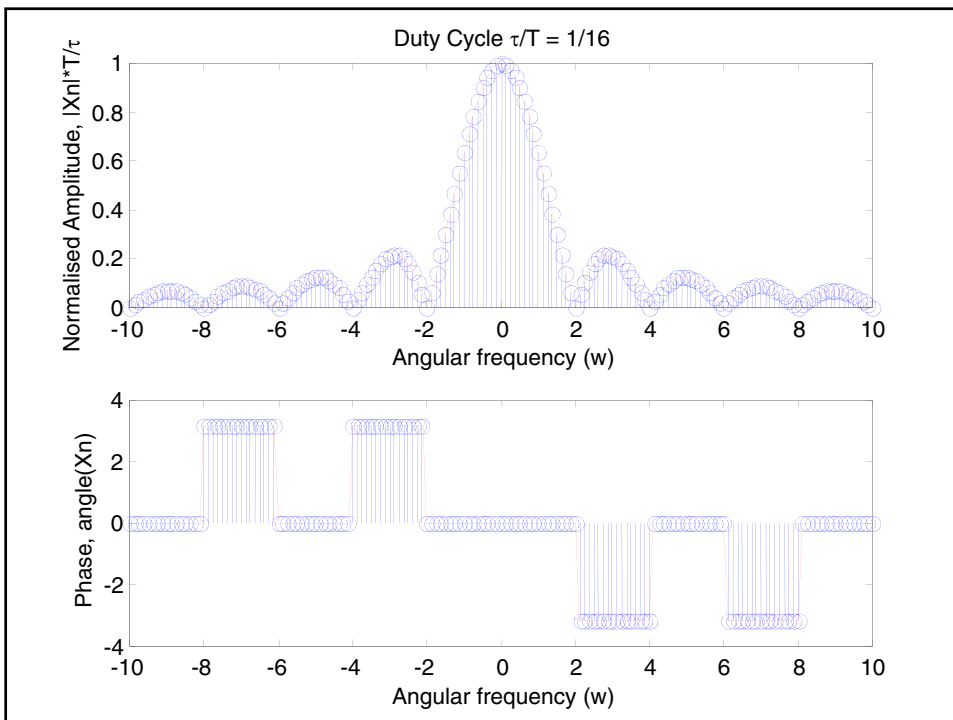
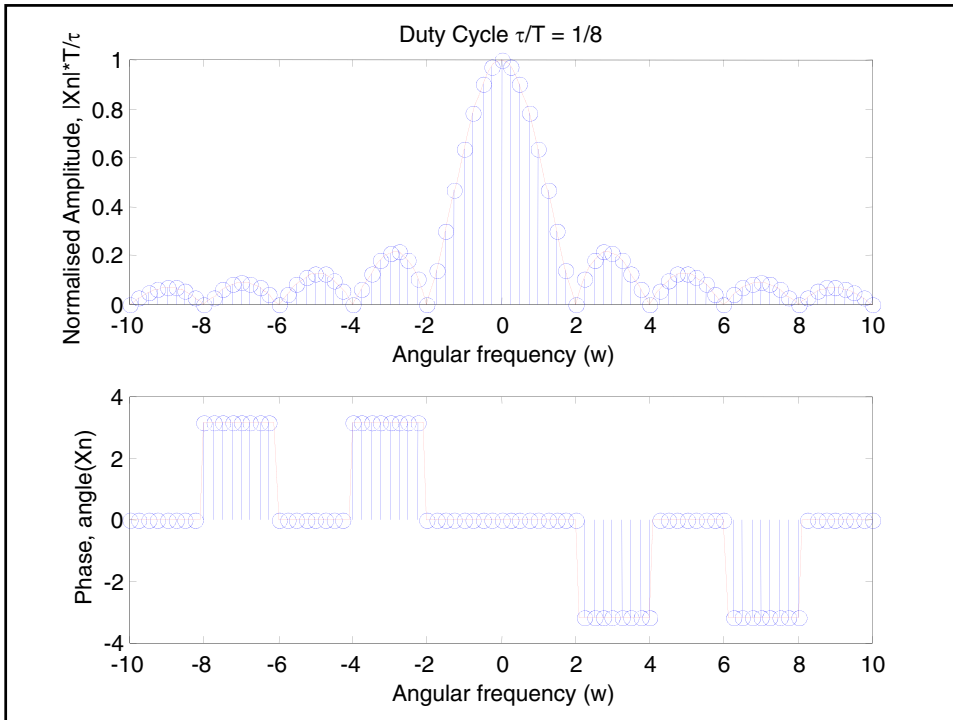
$$P = \frac{1}{T} \int_0^T x^2(t) dt = \sum_{n=-\infty}^{\infty} |X_n|^2$$

Continuous &
periodic

Discrete







Fourier Transform

- Fourier series
 - Only applicable to periodic signals
- Real world signals are rarely periodic
- Develop Fourier transform by
 - Examining a periodic signal
 - Extending the period to infinity



Fourier Transform

- Problem: as $T \rightarrow \infty$, $X_n \rightarrow 0$
 - i.e., Fourier coefficients vanish!
- Solution: re-define coefficients
 - $X_n' = T \times X_n$
- As $T \rightarrow \infty$
 - (harmonic frequency) $n\omega_0 \rightarrow \omega$ (continuous freq.)
 - (discrete spectrum) $X_n' \rightarrow X(\omega)$ (continuous spect.)
 - ω_0 (fundamental freq.) reduces $\rightarrow d\omega$ (differential)
 - Summation becomes integration



Fourier Transform

Note missing 1/T

$$X'_n = \int_{-T/2}^{+T/2} x(t) \exp(-jn\omega_0 t) dt \quad \text{Modified Fourier series}$$

$$X(\omega) = \lim_{T \rightarrow \infty} \int_{-T/2}^{+T/2} x(t) \exp(-jn\omega_0 t) dt \quad \text{Note: as } T \rightarrow \infty \\ n\omega_0 \rightarrow \omega$$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) \exp(-j\omega t) dt \quad \text{Fourier transform} \\ X(\omega) = F\{x(t)\}$$

$$x(t) = \lim_{T \rightarrow \infty} \sum_{-\infty}^{\infty} X'_n \exp(jn\omega_0 t) \frac{\omega_0}{2\pi} \quad \text{Inverse Fourier series}$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) \exp(j\omega t) d\omega \quad \text{Inverse Fourier transform} \\ x(t) = F^{-1}\{X(\omega)\}$$



As an example of the evaluation of the Fourier transform, consider the finite energy signal $x(t)$ illustrated in Figure 1.8.

$$\begin{aligned} X(\omega) &= \int_{-\infty}^{\infty} x(t) \exp(-j\omega t) dt \\ &= \int_{-\tau/2}^{\tau/2} x(t) \exp(-j\omega t) dt \\ &= \frac{-A}{j\omega} \left[\exp\left(\frac{-j\omega\tau}{2}\right) - \exp\left(\frac{j\omega\tau}{2}\right) \right] \\ &= A\tau \frac{\sin(\omega\tau/2)}{\omega\tau/2} \\ &= A\tau \operatorname{sinc}(\omega\tau/2) \end{aligned}$$

Using Euler's relation

$$\sin \theta = \frac{e^{j\theta} - e^{-j\theta}}{2j}$$

Note the similarity to Previous FS examples

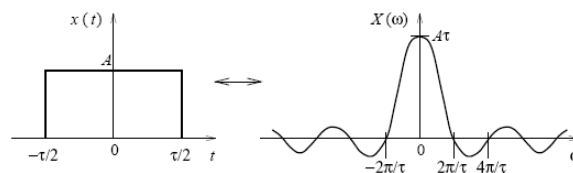


Figure 1.8 Fourier transform of a rectangular pulse.



Summary!

